



## MJE13002

## NPN EPITAXIAL SILICON TRANSISTOR

### HIGH VOLTAGE FAST-SWITCHING NPN POWER TRANSISTOR

#### DESCRIPTION

The UTC **MJE13002** designed for use in high-voltage, high speed, power switching in inductive circuit. It is particularly suited for 115 and 220V switchmode applications such as switching regulator's, inverters, DC-DC converter, Motor control, Solenoid/Relay drivers and deflection circuits.

#### FEATURES

\*Collector-Emitter Sustaining Voltage:

$V_{CEO(sus)} = 300V$ .

\*Collector-Emitter Saturation Voltage:

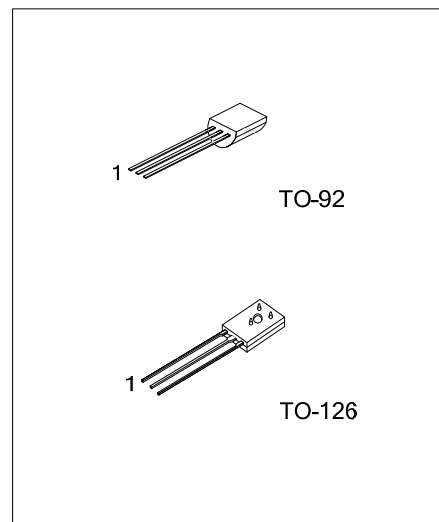
$V_{CE(sat)} = 1.0V(\text{Max.}) @ I_C = 1.0A, I_B = 0.25A$

\*Switch Time-  $t_f = 0.7\mu s(\text{Max.}) @ I_C = 1.0A$ .

#### ORDERING INFORMATION

Ordering Number		Package	Pin Assignment			Packing
Lead Free	Halogen Free		1	2	3	
MJE13002L-x-T92-B	MJE13002G-x-T92-B	TO-92	B	C	E	Tape Box
MJE13002L-x-T92-K	MJE13002G-x-T92-K	TO-92	B	C	E	Bulk
MJE13002L-x-T92-R	MJE13002G-x-T92-R	TO-92	B	C	E	Tape Reel
MJE13002L-x-T60-T	MJE13002G-x-T60-T	TO-126	B	C	E	Tube

MJE13002L-x-T92-K	(1) Packing Type (2) Package Type (3) Rank (4) Lead Free	(1) B: Tape Box, K: Bulk, R: Tape Reel, T: Tube (2) T92: TO-92, T60: TO-126 (3) x: refer to Classification of $h_{FE1}$ (4) G: Halogen Free, L: Lead Free
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## ■ ABSOLUTE MAXIMUM RATINGS

PARAMETER		SYMBOL	RATINGS	UNIT
Collector-Emitter Voltage		$V_{CEO(SUS)}$	300	V
Collector-Emitter Voltage		$V_{CEV}$	600	V
Emitter Base Voltage		$V_{EBO}$	9	V
Collector Current	Continuous	$I_C$	1.5	A
	Peak (1)	$I_{CM}$	3	
Base Current	Continuous	$I_B$	0.75	A
	Peak (1)	$I_{BM}$	1.5	
Emitter Current	Continuous	$I_E$	2.25	A
	Peak (1)	$I_{EM}$	4.5	
Total Power Dissipation	TA=25°C	$P_D$	1.4	Watts
	Derate above 25°C		11.2	MW/°C
	TC=25°C		40	Watts
	Derate above 25°C		320	MW/°C
Junction Temperature		$T_J$	150	°C
Storage Temperature		$T_{STG}$	-65 to +150	°C

## ■ THERMAL CHARACTERISTICS

PARAMETER		SYMBOL	RATINGS	UNIT
Junction to Case	TO-92	$\theta_{JC}$	25	°C/W
	TO-126		3.12	
Junction to Ambient	TO-92	$\theta_{JA}$	122	°C/W
	TO-126		89	
Maximum Load Temperature for Soldering Purposes: 1/8" from Case for 5 Seconds		$T_L$	275	°C

Note: 1. Pulse Test : Pulse Width=5ms,Duty Cycle≤10%

2. Designer 's Data for "Worst Case" Conditions – The Designer 's Data Sheet permits the design of most circuits entirely from the information presented. SOA Limit curves – representing boundaries on device characteristics – are given to facilitate "Worst case" design.

■ ELECTRICAL CHARACTERISTICS ( $T_C=25^\circ\text{C}$ , unless otherwise specified)

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT
OFF CHARACTERISTICS (1)						
Collector-Emitter Sustaining Voltage	V <sub>CEO(SUS)</sub>	I <sub>C</sub> =10 mA , I <sub>B</sub> =0	300			
Collector Cutoff Current	I <sub>CEV</sub>	V <sub>CEV</sub> =Rated Value, V <sub>BE(off)</sub> =1.5 V			1	
		V <sub>CEV</sub> =Rated Value, V <sub>BE(off)</sub> =1.5V,T <sub>C</sub> =100°C			5	
SECOND BREAKDOWN						
Second Breakdown Collector Current with bass forward biased (See Figure 5)	I <sub>S</sub> /I <sub>B</sub>					
Clamped Inductive SOA with base reverse biased (See Figure 6)	R <sub>BSOA</sub>					
DC Current Gain	h <sub>FE1</sub>	I <sub>C</sub> =0.5 A, V <sub>CE</sub> =2 V	8		40	
	h <sub>FE2</sub>	I <sub>C</sub> =1 A, V <sub>CE</sub> =2 V	5		25	
Collector-Emitter Saturation Voltage	V <sub>CE(SAT)</sub>	I <sub>C</sub> =0.5A, I <sub>B</sub> =0.1A			0.5	V
		I <sub>C</sub> =1A, I <sub>B</sub> =0.25A			1	
		I <sub>C</sub> =1.5A, I <sub>B</sub> =0.5A			3	
		I <sub>C</sub> =1A, I <sub>B</sub> =0.25A,T <sub>C</sub> =100°C			1	
Base-Emitter Saturation Voltage	V <sub>BE(SAT)</sub>	I <sub>C</sub> =0.5A, I <sub>B</sub> =0.1A			1	V
		I <sub>C</sub> =1A, I <sub>B</sub> =0.25 A			1.2	
		I <sub>C</sub> =1A, I <sub>B</sub> =0.25A,T <sub>C</sub> =100°C			1.1	
DYNAMIC CHARACTERISTICS						
Current-Gain-Bandwidth Product	f <sub>T</sub>	I <sub>C</sub> =100mA, V <sub>CE</sub> =10 V, f=1MHz	4	10		MHz
Output Capacitance	Cob	V <sub>CB</sub> =10V, I <sub>E</sub> =0, f=0.1MHz		21		pF
SWITCHING CHARACTERISTICS (TABLE 1)						
Delay Time	t <sub>d</sub>	V <sub>CC</sub> =125V, I <sub>C</sub> =1A, I <sub>B1</sub> =I <sub>B2</sub> =0.2A, t <sub>p</sub> =25μs, Duty Cycle≤1%		0.05	0.1	μs
Rise Time	t <sub>r</sub>			0.5	1	μs
Storage Time	t <sub>s</sub>			2	4	μs
Fall Time	t <sub>f</sub>			0.4	0.7	μs
INDUCTIVE LOAD, CLAMPED (TABLE 1, FIGURE 7)						
Storage Time	t <sub>sv</sub>	I <sub>C</sub> =1A,Vclamp=300V, I <sub>B1</sub> =0.2A,V <sub>BE</sub> (off)=5V,T <sub>C</sub> =100°C		1.7	4	μs
Crossover Time	t <sub>c</sub>			0.29	0.75	μs
Fall Time	t <sub>fi</sub>			0.15		μs

■ CLASSIFICATION OF  $h_{FE1}$

RANK	A	B	C	D	E	F
RANGE	8 ~ 16	15 ~ 21	20 ~ 26	25 ~ 31	30 ~ 36	35 ~ 40

## ■ APPLICATION INFORMATION

Table 1. Test Conditions for Dynamic Performance

[illegible]

Table 2. Typical Inductive Switching Performance

I <sub>C</sub> (AMP)	T <sub>C</sub> (°C)	T <sub>SV</sub> (μs)	T <sub>RV</sub> (μs)	T <sub>FI</sub> (μs)	T <sub>TI</sub> (μs)	T <sub>C</sub> (μs)
0.5	25	1.3	0.23	0.30	0.35	0.30
	100	1.6	0.26	0.30	0.40	0.36
1	25	1.5	0.10	0.14	0.05	0.16
	100	1.7	0.13	0.26	0.06	0.29
1.5	25	1.8	0.07	0.10	0.05	0.16
	100	3	0.08	0.22	0.08	0.28

Note: All Data Recorded in the inductive Switching Circuit Table 1

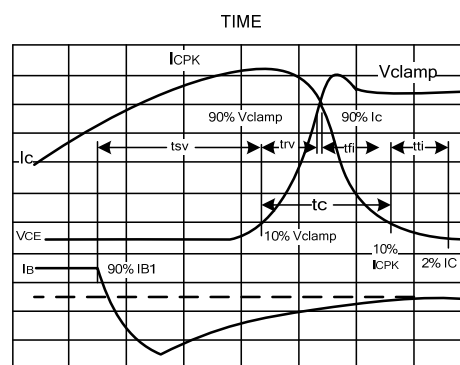


Fig 1. Inductive Switching Measurements

## SWITCHING TIMES NOTE

In resistive switching circuits, rise, fall, and storage times have been defined and apply to both current and voltage waveforms since they are in phase. However, for inductive loads which are common to SWITCHMODE power supplies and hammer drivers, current and voltage waveforms are not in phase. Therefore, separate measurements must be made on each wave form to determine the total switching time. For this reason, the following new terms have been defined.

$t_{SV}$ =Voltage Storage Time, 90%  $I_{B1}$  to 10%  $V_{clamp}$

$t_{RV}$ =Voltage Rise Time, 10-90%  $V_{clamp}$

$t_{FI}$ =Current Fall Time, 90-10%  $I_C$

$t_{TI}$ =Current Tail, 10-2%  $I_C$

$t_C$ =Crossover Time, 10%  $V_{clamp}$  to 10%  $I_C$

An enlarged portion of the inductive switching waveforms is shown in Figure 1 to aid in the visual identity of these terms.

For the designer, there is minimal switching loss during storage time and the predominant switching power losses occur during the crossover interval and can be obtained using the standard equation from AN-222:

$$PSWT = 1/2 V_{CC} I_C (t_C) f$$

In general,  $t_{RV} + t_{FI} \approx t_C$ . However, at lower test currents this relationship may not be valid.

As is common with most switching transistor, resistive switching is specified at 25°C and has become a benchmark for designers. However, for designers of high frequency converter circuits, the user oriented specifications which make this a "SWITCHMODE" transistor are the inductive switching speeds ( $t_C$  and  $t_{SV}$ ) which are guaranteed at 100°C.

## RESISTIVE SWITCHING PERFORMANCE

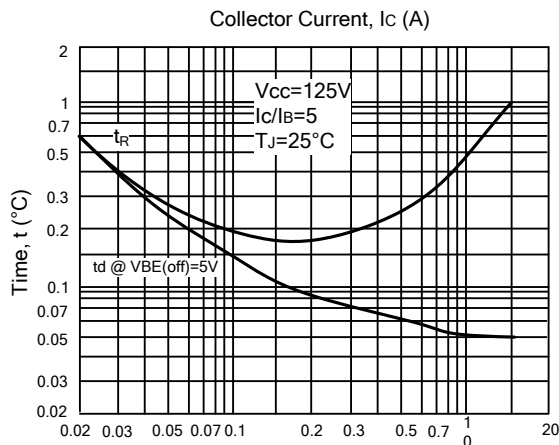


Fig 2. Turn-On Time

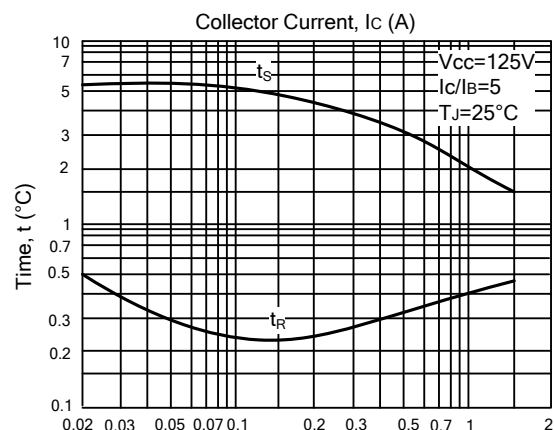


Fig 3. Turn-Off Time

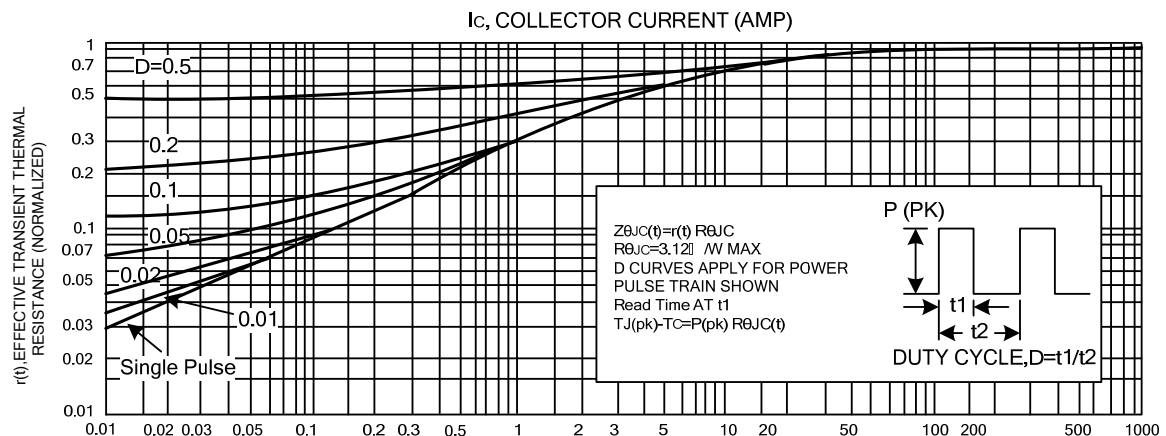


Fig 4. Thermal Response

## ■ SAFE OPERATING AREA INFORMATION

### FORWARD BIAS

There are two limitations on the power handling ability of a transistor: average junction temperature and second break-down. Safe operating area curves indicate  $I_C$  –  $V_{CE}$  limits of the transistor that must be observed for reliable operation; i.e., the transistor must not be subjected to greater dissipation than the curves indicate.

The data of Figure 5 is based on  $T_C=25^\circ\text{C}$ ;  $T_J(pk)$  is variable depending on power level. Second breakdown pulse limits are valid for duty cycles to 10% but must be derated when  $T_C \geq 25^\circ\text{C}$ . Second breakdown limitations do not derate the same as thermal limitations. Allowable current at the voltages shown on Figure 5 may be found at any case temperature by using the appropriate curve on Figure 7.

$T_J(pk)$  may be calculated from the data in Figure 5. At high case temperatures, thermal limitations will reduce the power that can be handled to values less than the limitations imposed by second breakdown.

### REVERSE BIAS

For inductive loads, high voltage and high current must be sustained simultaneously during turn-off, in most cases, with the base to emitter junction reverse biased. Under these conditions the collector voltage must be held to a safe level at or below a specific value of collector current. This can be accomplished by several means such as active clamping, RC snubbing, load line shaping, etc. The safe level for these devices is specified as Reverse Bias Safe Operating Area and represents the voltage-current conditions during re-verse biased turn-off. This rating is verified under clamped conditions so that the device is never subjected to an avalanche mode. Figure 6 gives RBSOA characteristics.

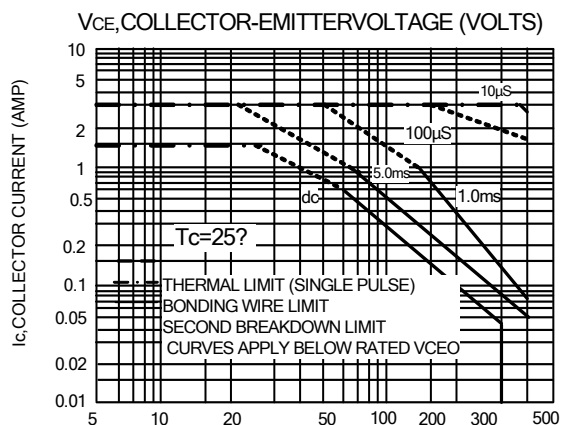


Fig 5. Active Region Safe Operating Area

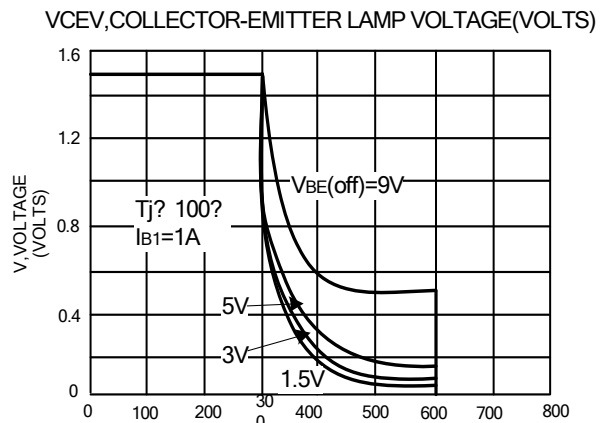


Fig 6. Reverse Bias Safe Operating Area

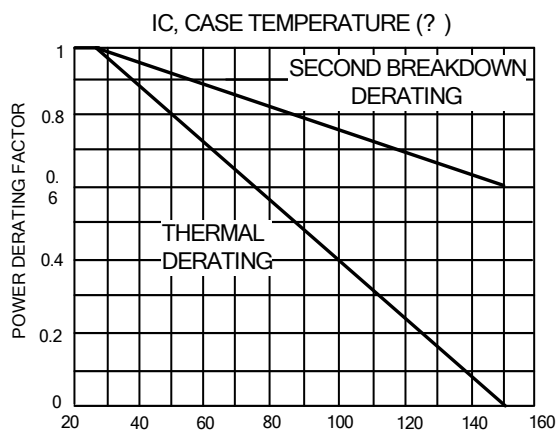
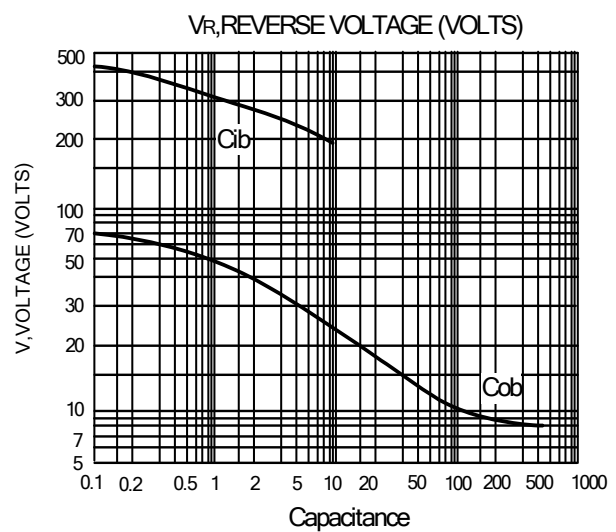
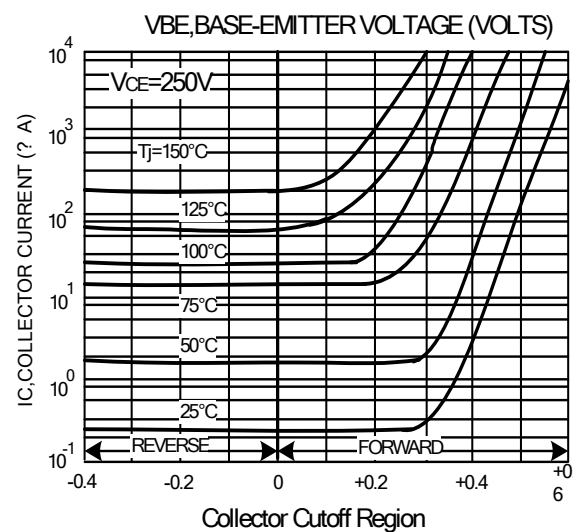
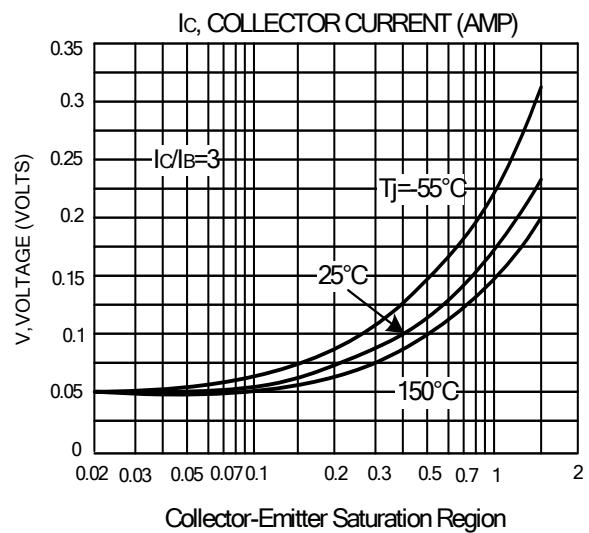
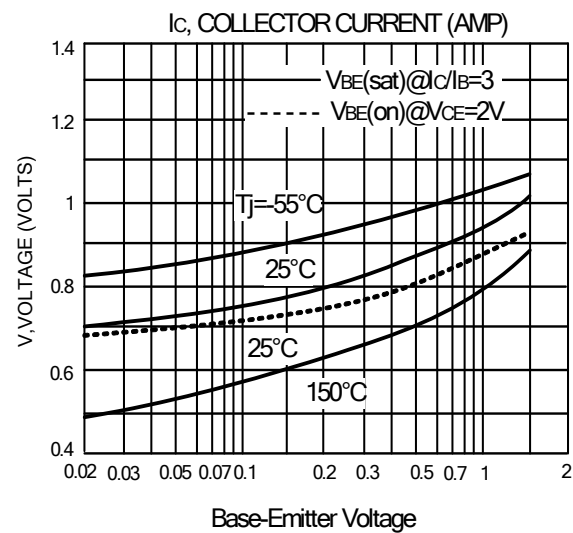
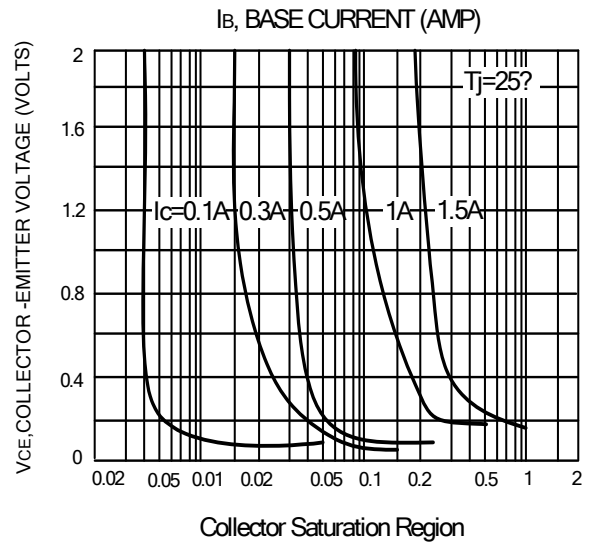
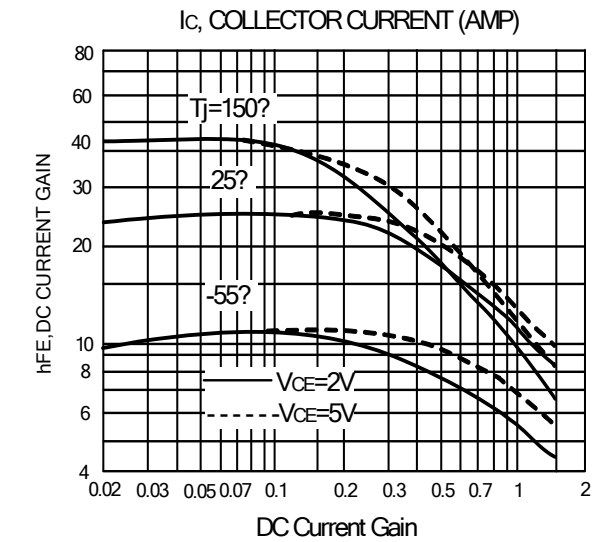


Fig 7. Forward Bias Power Derating

## ■ TYPICAL CHARACTERISTICS







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